

### Supplement S6.1 - Noise Margins for the Saturated Load Inverter

We will now calculate the values of the  $V_{IL}$ ,  $V_{OH}$ ,  $V_{IH}$  and  $V_{OL}$  for the inverter with a saturated load device. Remember that  $V_{IH}$  and  $V_{IL}$  are defined by the points in the transfer function at which the slope is equal to -1. In Fig. S6.1.1 the slope of the transfer function abruptly changes as  $M_S$  begins to conduct at the point where  $v_I = V_{TNS}$ . This point defines  $V_{IL}$ :

$$V_{IL} = V_{TNS} = 1 \text{ V} \quad \text{for} \quad V_{OH} = V_H = V_{DD} - V_{TNL} = 3.39 \text{ V}.$$

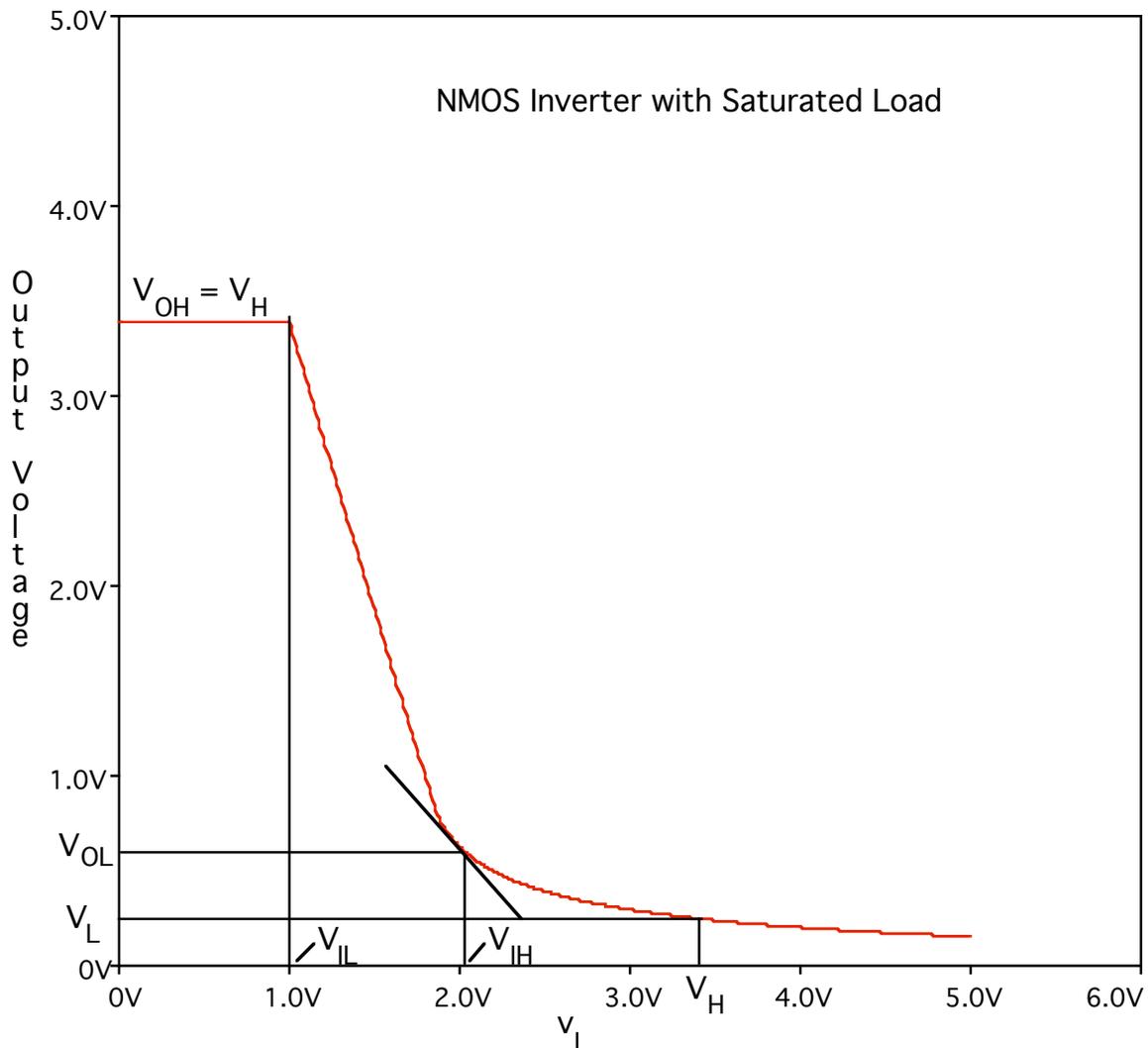


Figure S6.1.1 - PSPICE simulation of the voltage transfer function for the NMOS inverter with saturated load

Next let us find  $V_{IH}$ . In order to find a relationship between  $v_I$  and  $v_O$ , we observe that the drain currents in the switching and load devices must be equal. At  $v_I = V_{IH}$ , the input will be at a relatively high voltage, and the output will be at a relatively low voltage. Thus, we can guess that  $M_S$  will be in the linear region, and we already know that the circuit connection forces  $M_L$  to operate in the saturation region. Equating drain currents in the switching and load transistors:

$$i_{DS} = i_{DL}$$

$$K_S \left( v_I - V_{TNS} - \frac{v_O}{2} \right)^2 = \frac{K_L}{2} (V_{DD} - v_O - V_{TNL})^2$$

with

$$(S6.1.1)$$

$$K_S = K'_n \frac{W}{L}_S \quad \text{and} \quad K_L = K'_n \frac{W}{L}_L$$

The point of interest is  $\frac{\partial v_O}{\partial v_I} = -1$ , but solving for the value of  $v_O$  can be quite tedious.

Since the derivatives are smooth, continuous and non-zero, we will assume that

$\frac{\partial v_O}{\partial v_I} = -\frac{\partial v_I}{\partial v_O}$ , and will solve for  $v_I$  in terms of  $v_O$ :

$$v_I = V_{TNS} + \frac{v_O}{2} + \frac{1}{2K_R v_O} (V_{DD} - v_O - V_{TNL})^2 \quad \text{where} \quad K_R = \frac{K_S}{K_L}$$

or

$$v_I = V_{TNS} + \frac{v_O}{2} + \frac{1}{2K_R} \frac{(V_{DD} - V_{TNL})^2}{v_O} - 2(V_{DD} - V_{TNL}) + v_O$$

and

$$(S6.1.2)$$

$$\frac{\partial v_I}{\partial v_O} = \frac{1}{2} + \frac{1}{2K_R} \frac{(V_{DD} - V_{TNL})^2}{v_O^2} + 1$$

In this last expression, the dependence of  $V_{TNL}$  on  $v_O$  has been neglected for simplicity.

(This approximation will be justified shortly.)

Setting the derivative equal to -1 at  $v_I = V_{IH}$  yields

$$1 = \frac{1}{2} + \frac{1}{2K_R} \frac{(V_{DD} - V_{TNL})^2}{v_O^2} + 1$$

and solving for  $V_{OL} = v_O$  for  $v_I = V_{IH}$  yields:

$$V_{OL} = \frac{V_{DD} - V_{TNL}}{\sqrt{1 + 3K_R}} \quad \text{and} \quad V_{IH} = V_{TNS} + \frac{V_{OL}}{2} + \frac{(V_{DD} - V_{OL} - V_{TNL})^2}{2K_R V_{OL}} \quad (\text{S6.1.3})$$

For the inverter design of Fig. 6.26 with  $V_{TNL} = 1$  V, we find

$$V_{OL} = \frac{V_{DD} - V_{TNL}}{\sqrt{1 + 3K_R}} = \frac{V_{DD} - V_{TNL}}{\sqrt{1 + 3 \frac{(W/L)_S}{(W/L)_L}}} = \frac{(5-1)V}{\sqrt{1 + 3(3.53)(3.39)}} = 0.66V$$

$$V_{IH} = 1 + \frac{0.66V}{2} + \frac{1}{2(3.53)(3.39)} \frac{1}{0.66V} (5 - 0.66 - 1)^2 V^2 = 2.04V$$

With these values we can check our assumption of the operating region of  $M_S$ :

$$V_{GS} - V_{TNS} = 2.04 - 1 = 1.04 \text{ V} \quad \text{and} \quad V_{DS} = 0.66 \text{ V} .$$

Since  $V_{DS} < V_{GS} - V_{TN}$ , the linear region assumption was correct.

In Fig. S6.1.1, it can be seen that the calculated values of  $V_{IH}$  and  $V_{IL}$  agree well with SPICE simulation results. Thus, the approximation of neglecting the voltage dependence of the threshold voltage of the load device does not introduce significant error.

However, if we so desire, we can use an iterative procedure to improve on the estimates above. The calculated value of  $V_{OL}$  can be used to determine a better estimate for  $V_{TNL}$ , and this new value of  $V_{TNL}$  can then be used to improve the estimate of  $V_{IH}$ . For  $V_{OL} = 0.66V$ ,

$$V_{TNL} = 1V + 0.5\sqrt{V} \left( \sqrt{(0.66 + 0.6)V} - \sqrt{0.6V} \right) = 1.17 \text{ V}$$

and the new values of  $V_{OL}$  and  $V_{IH}$  from Eqn. (S6.1.3) are equal to

$$V_{OL} = 0.63 \text{ V} \quad \text{and} \quad V_{IH} = 1.99 \text{ V} .$$

This process can be continued, but it has already converged as indicated in Table 7.2 below. (Note that this procedure is another example of an algorithm that is very easy to implement using MATLAB or a spreadsheet.)

Table S6.1.1 - Iterative Update of $V_{OL}$ and $V_{IH}$				
Iteration #	$V_{OL}$	$V_{TNL}$	$V'_{OL}$	$V_{IH}$
0	---	1.00 V	0.66 V	2.04 V
1	0.66 V	1.17 V	0.63 V	1.99 V
2	0.63 V	1.17 V	0.63 V	1.99 V

$$NM_L = V_{IL} - V_{OL} = 1 - 0.63 = 0.37 \text{ V}$$

$$NM_H = V_{OH} - V_{IH} = 3.39 - 1.99 = 1.40 \text{ V}$$